



AMPLIFIER CIRCUITS
FOR KT77

The KT77 may be used in pairs in either triode, tetrode or ultra-linear push-pull circuits and this report gives circuits and operating data for amplifiers giving outputs of 25 W and 60 W in ultra-linear push-pull operation.

Either cathode bias or fixed bias may be used and the circuits show both types of operation. The former has the advantage of simplicity whereas the latter provides greater output and higher efficiency.

The output power of the ultra-linear circuit is not less than that of the tetrode at any given supply voltage and has the advantage that a low impedance screen supply is not required. The ultra-linear circuit does, in fact, show a rather higher efficiency in that a lower current is required from the power supply.

Owing to the high mutual conductance of the KT77 some precautions have to be taken against parasitic oscillation whatever the circuit arrangement. Grid and screen series resistors of about $10\text{ k}\Omega$ and $33\ \Omega$ respectively, are recommended.

A 25 W ULTRA-LINEAR AMPLIFIER USING CATHODE BIAS

This amplifier provides an output of 25 W with an ht supply of 400 V. The distortion due to the output stage is only 2 %, but the application of 20 dB negative feedback reduces the overall distortion of 0.2 %. The amplifier features low hum and noise, high sensitivity, high damping factor, wide frequency range and fast transient response.

The output stage consists of two KT77 tetrodes operating in Class AB1 with separate cathode resistors. This is preceded by a conventional double triode amplifier which is fed by a triode phase-splitter. The input stage voltage amplifier is directly coupled to the phase-splitter in order to minimise low frequency phase shift. The circuit diagram is shown in fig. 1.

Depending upon the particular characteristics of the output transformer, it may be necessary to prevent instability by connecting capacitors and resistors in series between the screen tap and the anode connection of each half-primary of the output transformer. Typical values are 1000 pF and $470\ \Omega$. The curves of fig. 2 show the distortion and intermodulation performance of the amplifier with and without feedback. The frequency response is shown in fig. 3.

TABLE 1 Component Values for Fig.1

Valves

| | | |
|----|-------|------------|
| V1 | | ECC83/B339 |
| V2 | | ECC83/B339 |
| V3 | | KT77 |
| V4 | | KT77 |
| V5 | | U77/GZ34 |

Resistors

(± 20 % 0.25 W unless otherwise stated)

| | | | | | |
|-----|-------|------------------|------------------|--------|-----------------------------|
| R1 | 1 MΩ | | R14 | 10 kΩ | 1 W |
| R2 | 68 kΩ | 0.5 W | R15 | 680 kΩ | } matched to 5 % |
| R3 | 470 Ω | | R16 | 680 kΩ | |
| R4 | 47 Ω | | R17 | 10 kΩ | |
| R5 | 47 kΩ | 0.5 W | R18 | 10 kΩ | |
| R6 | 15 kΩ | 1 W | R19 | 33 Ω | |
| R7 | 33 kΩ | 1 W | R20 | 33 Ω | |
| R8 | 33 kΩ | 1 W | R21 | 470 Ω | 5 W |
| R9 | 1 MΩ | } matched to 5 % | R22 | 470 Ω | 5 W |
| R10 | 1 MΩ | | } matched to 5 % | *R23 | 250 Ω |
| R11 | 390 Ω | 0.5 W | | R24 | 2000 √speech coil impedance |
| R12 | 47 kΩ | 1 W | R25 | 4.7 kΩ | |
| R13 | 47 kΩ | 1 W | | | |

*R23 Adjust to give ht line voltage of 400 V. This value, together with the resistance of half the secondary of T2 plus the reflected primary resistance of T2 must not be less than 250 Ω.

Capacitors

| | | | | | |
|----|--------------|-------|-----|--------------|---------------------|
| C1 | 8 μF | 350 V | C7 | 0.05 μF | 500 V |
| C2 | } 32 + 32 μF | 500 V | C8 | 50 μF | 50 V |
| C3 | | | C9 | 50 μF | 50 V |
| C4 | 0.05 μF | 350 V | C10 | } 16 + 16 μF | 500 V (600 V surge) |
| C5 | 0.05 μF | 350 V | C11 | | |
| C6 | 0.05 μF | 500 V | C12 | 133 pF | |
| | | | C13 | 220 pF | |

Miscellaneous Components

| | | | |
|----|---|----------------------------------|------------------------------------|
| L1 | Smoothing Choke | Inductance: | 10 H at 200 mA |
| | | Resistance: | 150 Ω |
| T1 | Ultralinear Output Transformer with 43 % Taps | Anode-anode impedance: | 6.6 k Ω |
| | | Primary inductance: | 200 H |
| | | Leakage inductance: | P - S: 4 mH |
| | | | $\frac{1}{2}$ P - screen tap: 4 mH |
| | | Primary dc resistance: | 70 + 70 Ω |
| T2 | Mains Transformer | Secondaries: | 410 - 0 - 410 V, 180 mA |
| | | | 6.3 V, 5 A |
| | | | 5 V, 3 A |

TABLE 2 DC Voltages Measured at the Points Shown in Fig. 1 (Under quiescent 'no signal' conditions)

| Point of Measurement | Voltage | DC Range of Avo Model 8 (50 μ A fsd) (V) |
|----------------------|---------|--|
| 1 | 400 | 1000 |
| 2 | 340 | 1000 |
| 3 | 242 | 1000 |
| 4 | 172 | 250 |
| 5 | 72 | 100 |
| 6 | 0.79 | 2.5 |
| 7 | 74 | 100 |
| 8 | 167 | 250 |
| 9 | 175 | 250 |
| 10 | 2.1 | 2.5 |
| 11 | 397 | 1000 |
| 12 | 30 | 100 |

TABLE 3 Operating Conditions for Output Stage

| | | |
|---------------------|----------|------------|
| V_b | 400 | V |
| $V_{a,g2}$ | 397 | V |
| $I_{a,g2(o)}$ | 2 x 60 | mA |
| $I_{a,g2(max sig)}$ | 2 x 65 | mA |
| $P_{a,g2(o)}$ | 2 x 24 | W |
| $P_{a,g2(max sig)}$ | 2 x 13.5 | W |
| R_k | 2 x 470 | Ω |
| $-V_g(approx)$ | 30 | V |
| P_L | 25 | W |
| $R_{L(a-a)}$ | 6.6 | k Ω |

Performance of the Complete Amplifier

| | | |
|--|------------|---------------|
| Feedback | 20 | dB |
| Rated output | 25 | W |
| Input for rated output | 120 | mV rms |
| Total harmonic distortion (25 W) | 0.2 | % |
| Intermodulation distortion (25 W) | 0.6 | % |
| Damping factor (15 Ω load) | 16.5 | — |
| 3 dB frequency response (at 25 W = 0 dB) | 25 – 25000 | Hz |
| Hum (amplifier input open-circuit) | 65 | dB below 25 W |
| Noise (amplifier input open-circuit) | 72 | dB below 25 W |
| Hum (amplifier input short-circuit) | 76 | dB below 25 W |
| Noise (amplifier input short-circuit) | > 80 | dB below 25 W |
| Transient response rise time (approx) | 7.4 | μ s |

A 60 W ULTRA-LINEAR AMPLIFIER USING FIXED BIAS

This amplifier provides an output of 60 W with an ht supply of 550 V. The distortion due to the output stage is 3 % but this is reduced to 0.3 % when feedback is applied. The wide frequency response, fast transient response and low hum and noise are features of this amplifier.

The output stage consists of two KT77 tetrodes operating in Class AB1 with separate fixed bias adjustment for each valve. The bias is obtained from a separate rectifier supplied by a tap on the ht transformer. The driver stage is a double triode amplifier which is fed by a triode phase-splitter. The circuit diagram is shown in fig. 4.

As in the 25 W amplifier it may be necessary to prevent instability by connecting capacitors and resistors in series between the screen tap and the anode connection of each half-primary of the output transformer. Typical values are 1000 pF and 470 Ω . The curves of figs. 5 and 6 show the distortion and intermodulation performance of the amplifier with and without feedback. Fig. 7 shows the frequency response characteristics.

TABLE 4 Component Values for Fig. 4

Valves

| | |
|--------------|-------------------------------|
| V1 | ECC82/B329 |
| V2 | ECC82/B329 |
| V3 | KT77 |
| V4 | KT77 |
| V5 | U19 or GXU50 |
| V6 | U19 or GXU50 |
| V7 | S102/2K or alternative device |

Resistors

(± 20 % 0.25 W unless otherwise stated)

| | | | | | |
|-----|--------|------|-----|--------|--------------------|
| R1 | 1 MΩ | | R15 | 10 kΩ | 5 W |
| R2 | 4.7 kΩ | | R16 | 220 kΩ | |
| R3 | 470 Ω | 10 % | R17 | 220 kΩ | |
| R4 | 33 kΩ | 1 W | R18 | 10 kΩ | |
| R5 | 47 kΩ | 1 W | R19 | 10 kΩ | |
| R6 | 47 Ω | | R20 | 33 Ω | |
| R7 | 22 kΩ | 1 W | R21 | 33 Ω | |
| R8 | 22 kΩ | 1 W | R22 | 250 | √speaker impedance |
| R9 | 22 kΩ | 1 W | R23 | 3.9 kΩ | 1 W |
| R10 | 470 kΩ | | R24 | 20 kΩ | W/W |
| R11 | 470 kΩ | | R25 | 20 kΩ | W/W |
| R12 | 680 Ω | | R26 | 100 kΩ | 1 W |
| R13 | 47 kΩ | 1 W | R27 | 100 kΩ | 1 W |
| R14 | 47 kΩ | 1 W | | | |

Capacitors

| | | | | | |
|----|---------|-------|-----|---------|---------------------|
| C1 | 8 μF | 350 V | C7 | 0.25 μF | 350 V |
| C2 | 220 pF | | C8 | 0.25 μF | 350 V |
| C3 | 0.05 μF | 350 V | C9 | 64 μF | 450 V (525 V surge) |
| C4 | 0.05 μF | 350 V | C10 | 64 μF | 450 V (525 V surge) |
| C5 | 16 μF | 500 V | C11 | 440 pF | |
| C6 | 16 μF | 500 V | C12 | 50 μF | 200 V |
| | | | C13 | 50 μF | 200 V |

Miscellaneous Components

| | | | | |
|-----|--|--|------------------------------|--|
| L1 | Smoothing Choke | | | |
| | Inductance : | | 10 H at 200 mA | |
| | dc resistance : | | 150 Ω | |
| T1 | Ultra-linear Output Transformer with 43 % Taps | | | |
| | Anode-anode impedance : | | 5.5 kΩ | |
| | Primary inductance : | | 4 28 H | |
| | Leakage inductance : | | P – S > 6 mH | |
| | | | ½P – screen tap > 6 mH | |
| | Primary dc resistance : | | 70 + 70 Ω | |
| T2 | Mains Transformer | | | |
| | Primary : | | 0 – 200 – 220 – 240 V | |
| | Secondary : | | 640 – 0 – 100 – 640 V 250 mA | |
| T3 | Filament Transformer | | | |
| | Secondaries : | | 6.3 V 5 A CT | |
| | | | 4.0 V 3.5 A | |
| MR1 | Sentercell contact rectifier Type C.2 H D.59 or silicone diode BYX10 etc. | | | |
| F1 | Mains Fuse | | 2 A | |
| F2 | HT Fuse | | 250 mA | |

**TABLE 5 DC Voltages Measured at the Points Shown on Fig. 4
(Under quiescent 'no signal' conditions)**

| Point of Measurement (Fig. 4) | Voltage (V) | DC Range of Avo Model 8 (50 μ A fsd) (V) |
|-------------------------------|-------------|--|
| 1 | 553 | 1000 |
| 2 | 389 | 1000 |
| 3 | 300 | 1000 |
| 4 | 263 | 1000 |
| 5 | 84 | 100 |
| 6 | 209 | 250 |
| 7 | 1.93 | 2.5 |
| 8 | 88 | 100 |
| 9 | 182 | 250 |
| 10 | 182 | 250 |
| 11 | 6.7 | 10 |
| 12 | 550 | 1000 |
| 13 | 550 | 1000 |

**TABLE 6 Operating Conditions for Output Stage
(Continuous Signal 1 kHz)**

| | | |
|----------------------|------|------------|
| V_b | 553 | V |
| $V_{a,g2(o)}$ | 550 | V |
| $V_{a,g2(max\ sig)}$ | 500 | V |
| $I_{a,g2(o)}$ | 45 | mA |
| $I_{a,g2(max\ sig)}$ | 110 | mA |
| $P_{a,g2(o)}$ | 25 | W |
| $P_{a,g2(max\ sig)}$ | 25 | W |
| $-V_g$ (approx) | 48 | V |
| P_L | 60 | W |
| $R_L(a-a)$ | 5.5 | k Ω |
| D_{tot} | 0.3 | % |
| Z_{out} | 1.5 | k Ω |
| *IM | 1.4 | % |
| Damping factor | 14.5 | - |

If negative feedback is omitted, the following changes in values occur:-

| | | |
|-----------|-----|------------|
| D_{tot} | 3.0 | % |
| Z_{out} | 20 | k Ω |
| *IM | 11 | % |

*Intermodulation distortion: measured using two input signals at 50 and 6000 Hz (ratio of amplitudes 4:1).

Performance of the Complete Amplifier

| | | |
|--|-----------|---------------|
| Feedback | 20 | dB |
| Rated output | 60 | W |
| Input for rated output | 1.65 | V rms |
| Total harmonic distortion | 0.3 | % |
| Intermodulation distortion | 1.4 | % |
| 3 dB frequency response (at 60 W = 0 dB). | < 20 – 45 | kHz |
| Hum and noise (source resistance 47 k Ω) | 74 | dB below 60 W |
| Transient response rise time | 1.5 | μ s |

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Fig. 2 INTERMODULATION AND TOTAL DISTORTION CURVES FOR 25 W AMPLIFIER

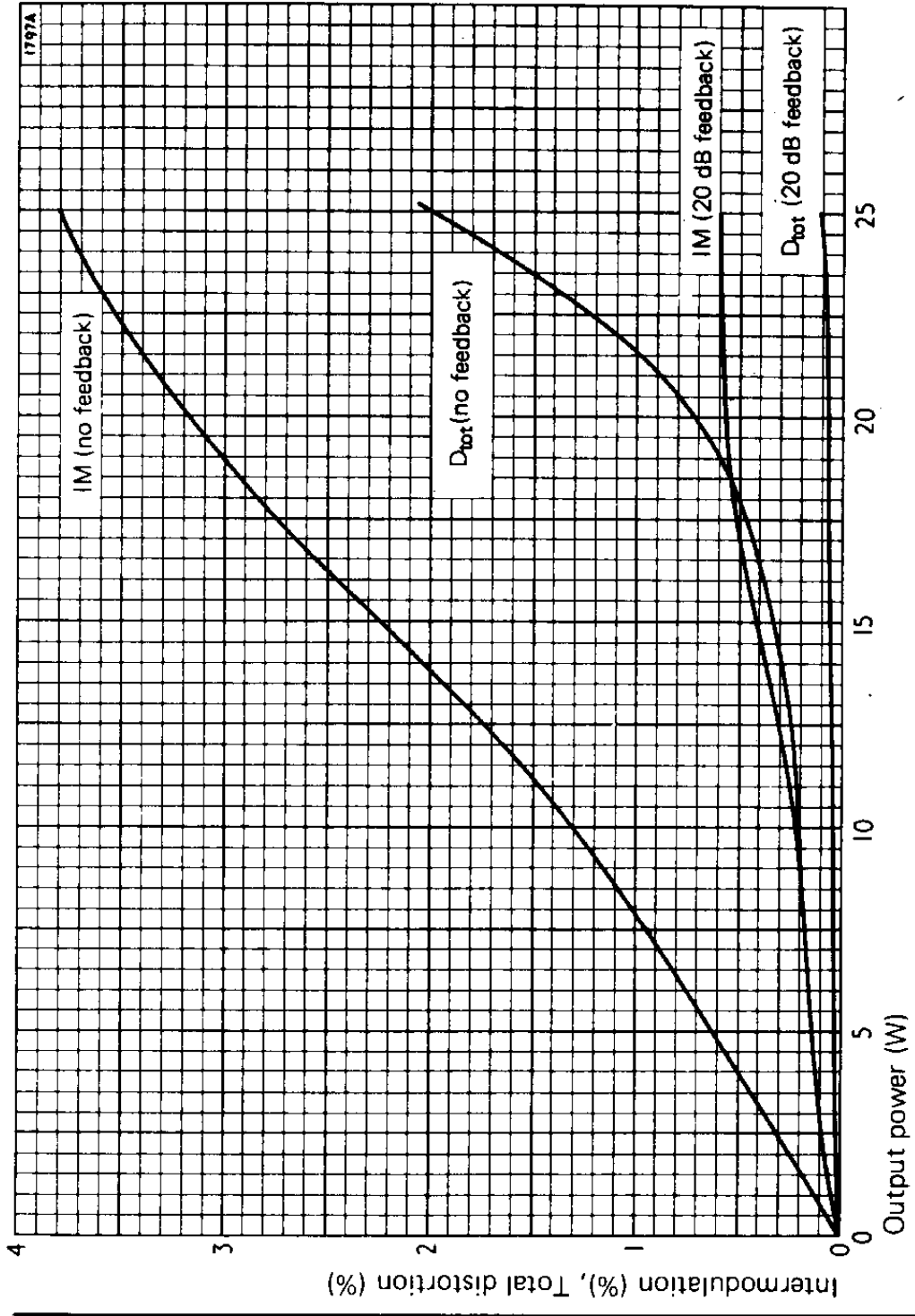


Fig. 3 FREQUENCY RESPONSE OF 25 W AMPLIFIER (20 dB FEED-BACK)

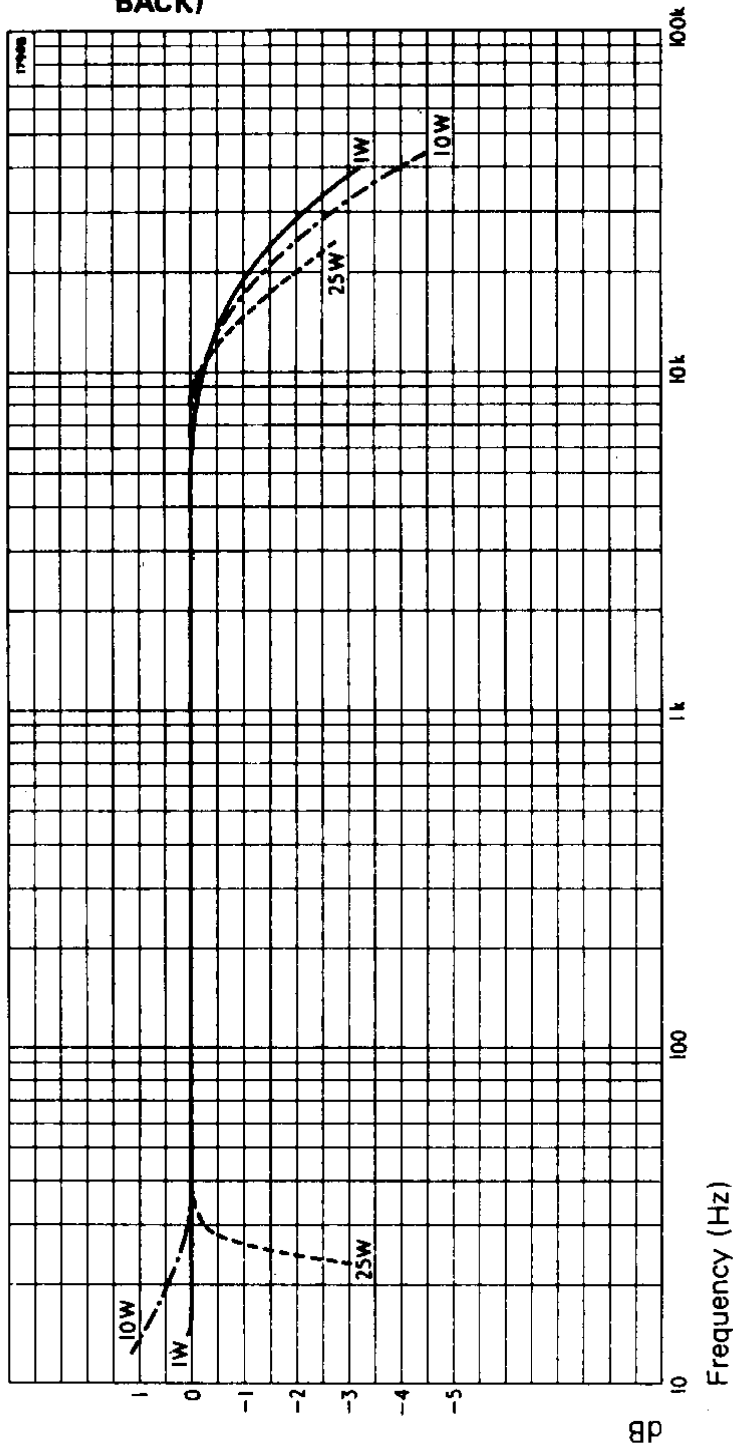
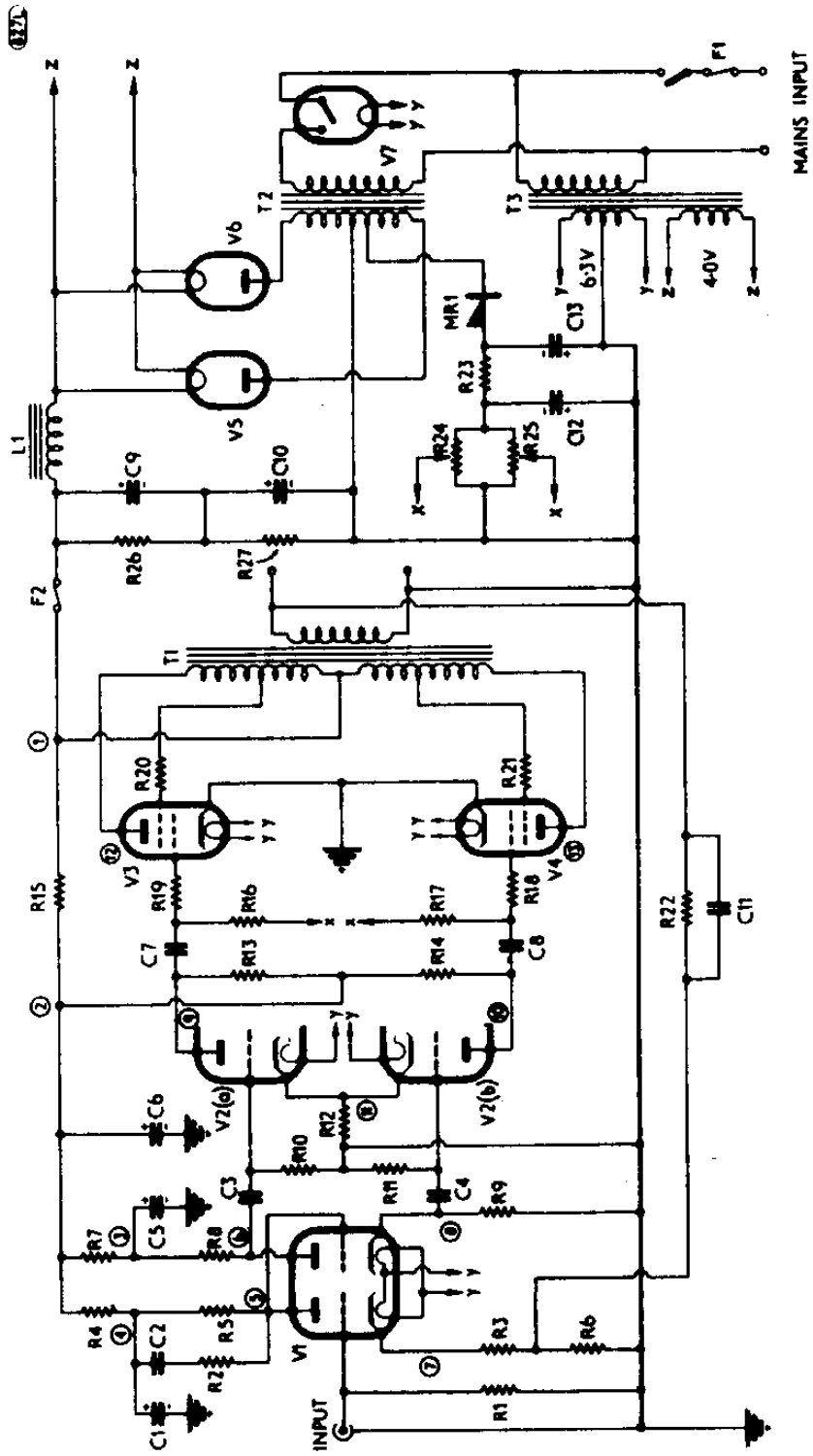


Fig. 4 CIRCUIT DIAGRAM OF 60 W ULTRA-LINEAR AMPLIFIER



Numbers in circles indicate measurement points shown in table 5

Fig. 5 TOTAL DISTORTION CURVES FOR 60 W AMPLIFIER

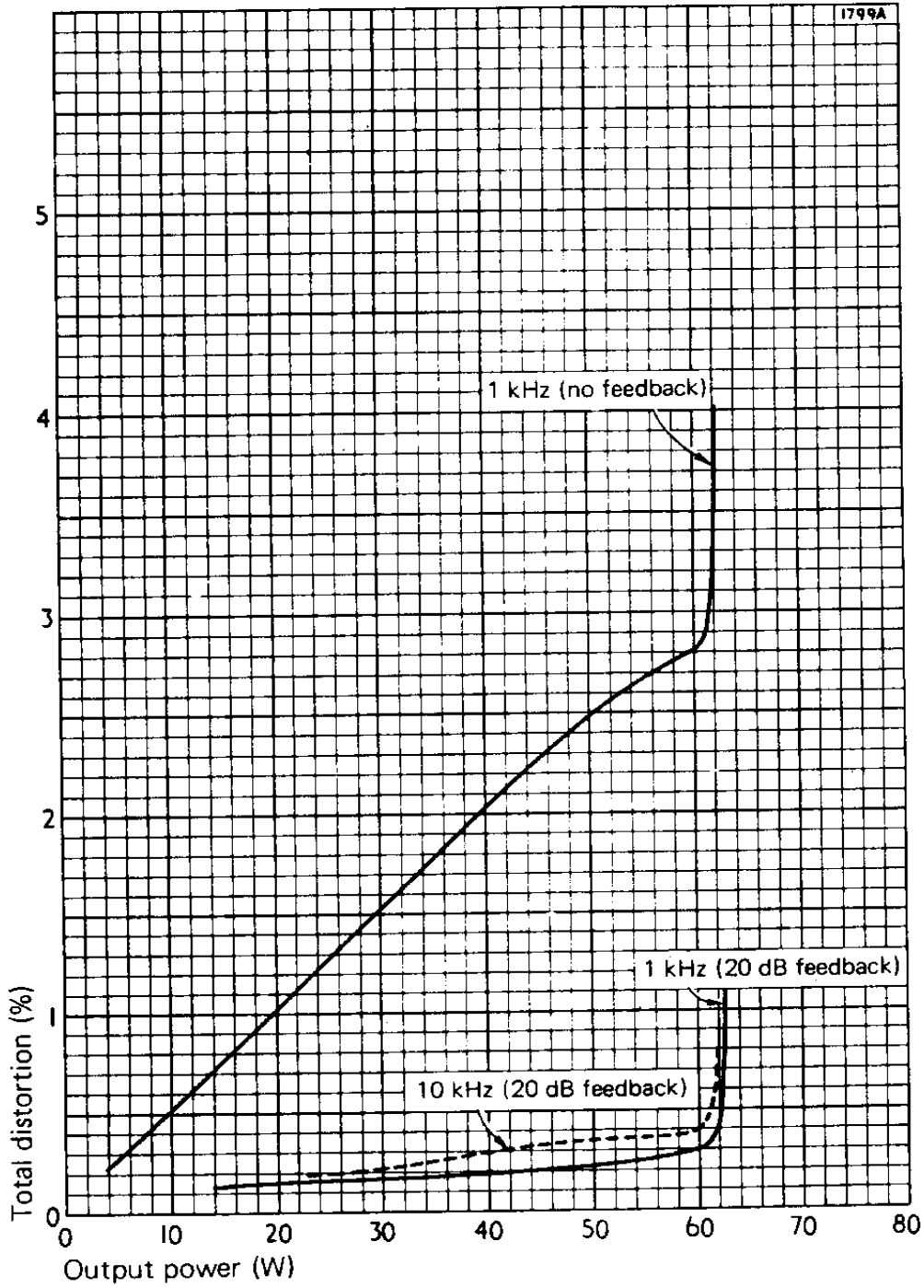
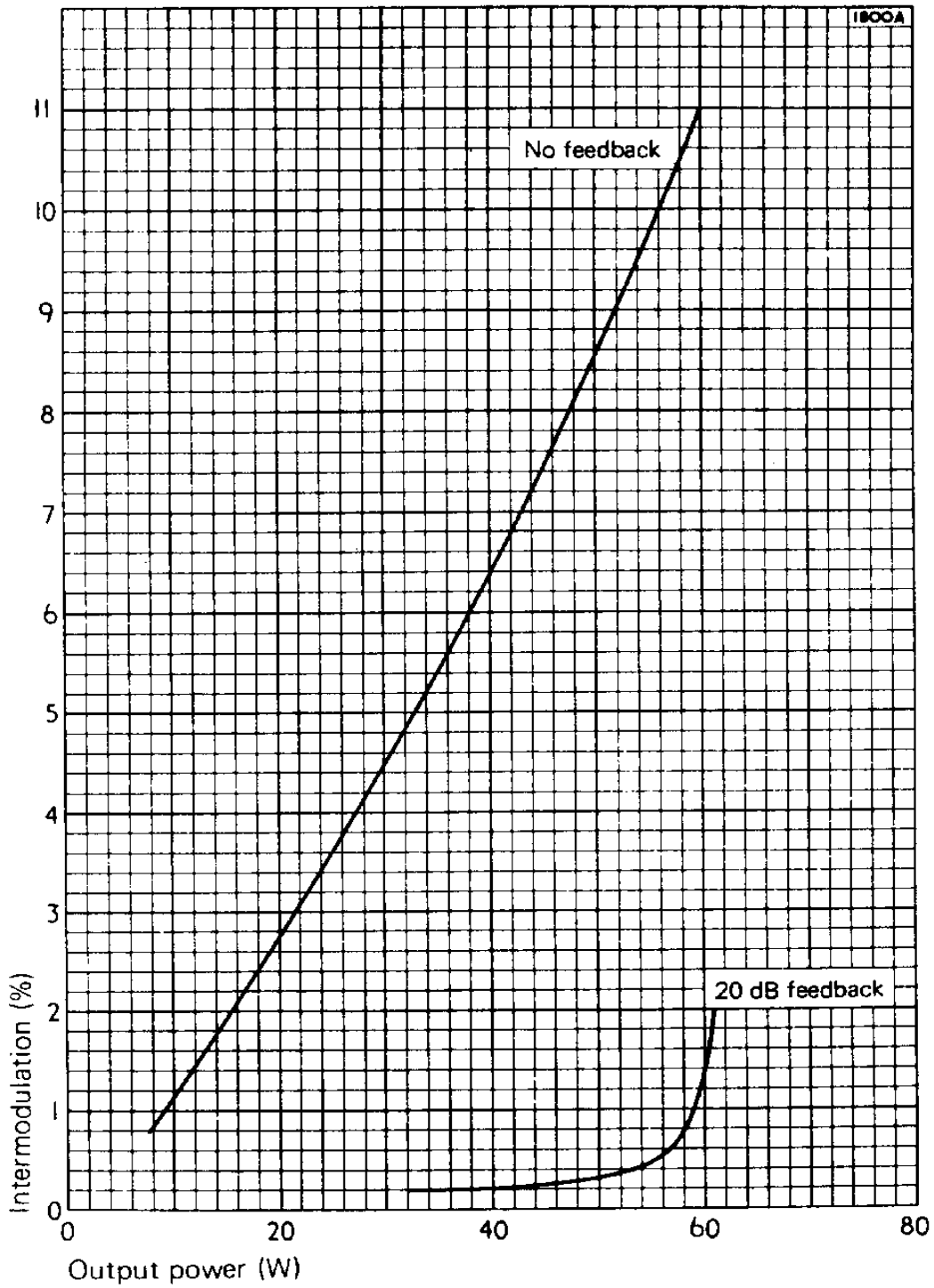


Fig. 6 INTERMODULATION DISTORTION CURVES FOR 60 W AMPLIFIER



**Fig. 7 FULL OUTPUT FREQUENCY RESPONSE OF 60 W AMPLIFIER
(20 dB FEEDBACK)**

